# Level Crossing Rate of the SSC Combiner Output Signal in the Presence of Log-normal Fading

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Abstract - Level crossing rate of the SSC combiner output signal are determined in this paper. The presence of the lognormal fading at the input is observed. The results are shown graphically for different variance values and decision threshold values.

Keywords - Level Crossing Rate, Log-normal Fading, SSC Combining

# I. INTRODUCTION

Many of the wireless communication systems use some form of diversity combining to reduce multupath fading appeared in the channel. Among the simpler diversity combining schemes, the two most popular are selection combining (SC) and switch and stay combining (SSC). SSC is an attempt at simplifying the complexity of the system but with loss in performance. In this case the receiver selects a particular antenna until its quality drops below a predetermined threshold. When this happens, the receiver switches to another antenna and stays with it for the next time slot, regardless of whether or not the channel quality of that antenna is above or below the predetermined threshold.

In the paper [1] Alouini and Simon develop, analyze and optimize a simple form of dual-branch switch and stay combining (SSC).

The consideration of SSC systems in the literature has been restricted to low-complexity mobile units where the number of diversity antennas is typically limited to two ([2], [3] and [4]). Furthermore, in all these publications, only predetection SSC has thus far been considered wherein the switching of the receiver between the two receiving antennas is based on a comparison of the instantaneous SNR of the connected antenna with a predetermined threshold. This results in a reduction in complexity relative to SC in that the simultaneous and continuous monitoring of both branches SNRs is no longer necessary.

In [5] the moment generating function (MGF) of the signal power at the output of dual-branch switch-and-stay selection

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<sup>4</sup>Ilija Temelkovski is with the Faculty of Electronic Engineering, A. Medvedeva 14, 18000 Nis, Serbia. diversity (SSC) combiners is derived.

In this paper level crossing rate of the SSC combiner output signal in the presence of log-normal fading will be determine. The results will be shown graphically for different variance values and decision threshold values.

#### II. SYSTEM MODEL

The model of the SSC combiner with two inputs, considered in this paper, is shown in Fig. 1. The signals at the combiner input are  $r_1$  and  $r_2$ , and r is the combiner output signal. The predetection combining is observed.



Fig. 1. Model of the SSC combiner with two inputs

The probability of the event that the combiner first examines the signal at the first input is  $P_1$ , and for the second input is  $P_2$ . If the combiner examines first the signal at the first input and if the value of the signal at the first input is above the treshold,  $r_T$ , SSC combiner forwards this signal to the circuit for the decision. If the value of the signal at the first input is below the treshold  $r_T$ , SSC combiner forwards the signal at the signal at the first input is below the treshold  $r_T$ , SSC combiner forwards the signal to the signal from the other input to the circuit for the decision.

We can select decision threshold value so that one of three parameters has to be minimal: the error probability, fade duration or average signal value.

If the SSC combiner first examines the signal from the second combiner input it works in the similar way.

The determination of the probability density of the combiner output signal is important for the receiver performances determination. The probability for the first input to be examined first is  $P_1$  and for the second input to be examined first is  $P_2$ .

## **III. SYSTEM PERFORMANCES**

The probability densitie functions (PDFs) of the combiner input signals,  $r_1$  and  $r_2$ , in the presence of log-normal fading, are:

$$p_{r_1}(r_1) = \frac{1}{\sqrt{2\pi}\sigma_1 r_1} e^{-\frac{(\ln r_1 - \mu_1)^2}{2\sigma_1^2}} \qquad r_1 \ge 0 \qquad (1)$$

$$p_{r_2}(r_2) = \frac{1}{\sqrt{2\pi}\sigma_2 r_2} e^{-\frac{(\ln r_2 - \mu_2)^2}{2\sigma_2^2}} \quad r_2 \ge 0 \quad (2)$$

The cumulative probability densities (CDFs) are given by:

$$F_{r_{1}}(r_{T}) = \int_{0}^{r_{T}} p_{r_{1}}(x) dx$$
(3)

$$F_{r_2}(r_T) = \int_{0}^{r_T} p_{r_2}(x) dx$$
(4)

 $r_T$  is the treshold of the decision. In the presence of lognormal fading CDFs are:

$$F_{r_{1}}(r_{T}) = \int_{0}^{r_{T}} \frac{1}{\sqrt{2\pi\sigma_{1}x}} e^{-\frac{(\ln x - \mu_{1})^{2}}{2\sigma_{1}^{2}}} dx = \frac{1}{2} + \frac{1}{2} \operatorname{erf}\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)$$
(5)  
$$F_{r_{2}}(r_{T}) = \int_{0}^{r_{T}} \frac{1}{\sqrt{2\pi\sigma_{2}x}} e^{-\frac{(\ln x - \mu_{2})^{2}}{2\sigma_{2}^{2}}} dx = \frac{1}{2} + \frac{1}{2} \operatorname{erf}\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)$$
(6)

where erfc(x) is the error function and it is defined as [7]:

$$erf(x) = \frac{2}{\sqrt{\pi}} \int_{0}^{x} e^{-t^{2}} dt$$

The joint probability densities of the combiner input signals,  $r_1$  and  $r_2$ , and their derivatives  $\dot{r}_1$  and  $\dot{r}_2$ , in the presence of log-normal fading, are:

$$p_{r_{1}\dot{r}_{1}}(r_{1},\dot{r}_{1}) = \frac{1}{\sqrt{2\pi}\sigma_{1}r_{1}}e^{-\frac{(\ln r_{1}-\mu_{1})^{2}}{2\sigma_{1}^{2}}} \cdot \frac{1}{\sqrt{2\pi}\beta_{1}r_{1}}e^{-\frac{\dot{r}_{1}^{2}}{2\beta_{1}^{2}r_{1}^{2}}}$$

$$r_{1} \ge 0 \qquad (7)$$

$$p_{r_{2}\dot{r}_{2}}(r_{2},\dot{r}_{2}) = \frac{1}{\sqrt{2\pi}\sigma_{2}r_{2}}e^{-\frac{(\ln r_{2}-\mu_{2})^{2}}{2\sigma_{2}^{2}}} \cdot \frac{1}{\sqrt{2\pi}\beta_{2}r_{2}}e^{-\frac{\dot{r}_{2}^{2}}{2\beta_{2}^{2}r_{2}^{2}}}$$

$$r_{2} \ge 0 \qquad (8)$$

The probabilities  $P_1$  and  $P_2$  are:

$$P_{1} = \frac{F_{r_{2}}(r_{T})}{F_{r_{1}}(r_{T}) + F_{r_{2}}(r_{T})} =$$

$$= \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \qquad (9)$$

$$P_{2} = \frac{F_{r_{1}}(r_{T})}{F_{r_{1}}(r_{T}) + F_{r_{2}}(r_{T})} =$$

$$= \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \qquad (10)$$

The expression for the joint probability density function of the SSC combiner output signal and its derivative will be determined first for the case:  $r < r_T$ :

$$p_{r\dot{r}}(r\dot{r}) = P_1 \cdot F_{r_1}(r_T) \cdot p_{r_2\dot{r}_2}(r\dot{r}) + P_2 \cdot F_{r_2}(r_T) \cdot p_{\dot{r}_1\dot{r}_1}(r\dot{r}) \quad (11)$$
  
and then for  $r \ge r_T$ :

$$p_{r\dot{r}}(r\dot{r}) = P_1 \cdot p_{r_1\dot{r}_1}(r\dot{r}) + P_1 \cdot F_{r_1}(r_T) \cdot p_{r_2\dot{r}_2}(r\dot{r}) + + P_2 \cdot p_{r_2\dot{r}_2}(r\dot{r}) + P_2 \cdot F_{r_2}(r_T) \cdot p_{r_1\dot{r}_1}(r\dot{r})$$
(12)

We have now for  $r < r_T$ :

$$p_{r\dot{r}}(r,\dot{r}) = \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \cdot \left(\frac{1}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi}\sigma_{2}r}e^{-\frac{(\ln r - \mu_{2})^{2}}{2\sigma_{2}^{2}}} \frac{1}{\sqrt{2\pi}\beta_{2}r}e^{-\frac{\dot{r}^{2}}{2\beta_{2}^{2}r^{2}}} + \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \cdot \frac{1}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{\sqrt{2\pi}\sigma_{1}r}e^{-\frac{(\ln r - \mu_{1})^{2}}{2\sigma_{1}^{2}}} \frac{1}{\sqrt{2\pi}\beta_{1}r}e^{-\frac{\dot{r}^{2}}{2\beta_{1}^{2}r^{2}}}$$
(13)

and for  $r \ge r_T$ :

$$\begin{split} p_{r\bar{r}}(r,\dot{r}) &= \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right) \sqrt{2\pi\sigma_{1}r}} e^{-\frac{(\ln r - \mu_{1})^{2}}{2\sigma_{1}^{2}}} \\ &\cdot \frac{1}{\sqrt{2\pi}\beta_{1}r} e^{-\frac{\dot{r}^{2}}{2\beta_{1}^{2}r^{2}}} + \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \\ &\cdot \left(\frac{1}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi\sigma_{2}r}} e^{-\frac{(\ln r - \mu_{2})^{2}}{2\sigma_{2}^{2}}} \frac{1}{\sqrt{2\pi}\beta_{2}r} e^{-\frac{\dot{r}^{2}}{2\beta_{2}^{2}r^{2}}} + \\ &+ \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)} + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \\ &\cdot \frac{1}{\sqrt{2\pi}\sigma_{2}r} e^{-\frac{(\ln r - \mu_{2})^{2}}{2\sigma_{2}^{2}}} \frac{1}{\sqrt{2\pi}\beta_{2}r} e^{-\frac{\dot{r}^{2}}{2\beta_{2}^{2}r^{2}}} + \\ &+ \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)} \\ &\cdot \frac{1}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \left(\frac{1}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)\right)}. \end{split}$$

$$\cdot \frac{1}{\sqrt{2\pi}\sigma_{1}r}e^{-\frac{(\ln r-\mu_{1})^{2}}{2\sigma_{1}^{2}}}\frac{1}{\sqrt{2\pi}\beta_{1}r}e^{-\frac{\dot{r}^{2}}{2\beta_{1}^{2}r^{2}}}$$
(14)

For the channels with identical parameters it is, for  $r < r_T$ :

$$p_{r\dot{r}}(r,\dot{r}) = \left(\frac{1}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu}{\sigma\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi}\sigma r} e^{-\frac{(\ln r - \mu)^{2}}{2\sigma^{2}}} \frac{1}{\sqrt{2\pi}\beta r} e^{-\frac{r^{2}}{2\beta^{2}r^{2}}}$$
(15)

and for  $r \ge r_T$ :

$$p_{r\dot{r}}(r,\dot{r}) = \left(\frac{3}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu}{\sigma\sqrt{2}}\right)\right)\frac{1}{\sqrt{2\pi}\sigma r}e^{-\frac{(\ln r - \mu)^{2}}{2\sigma^{2}}}\frac{1}{\sqrt{2\pi}\beta r}e^{-\frac{\dot{r}^{2}}{2\beta^{2}r^{2}}}$$
(16)

The level crossing rate is:

$$N(r_{th}) = \int_{0}^{\infty} \dot{r} \, p_{r\dot{r}}(r_{th}, \dot{r}) \, d\dot{r}$$
(17)

for  $r_{th} < r_T$ :

$$N(r_{th}) = \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \cdot \left(\frac{1}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi}\sigma_{2}r_{th}} e^{-\frac{(\ln r_{th} - \mu_{2})^{2}}{2\sigma_{2}^{2}}} \frac{\beta_{2}r_{th}}{\sqrt{2\pi}} + \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \cdot \left(\frac{1}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi}\sigma_{1}r_{th}} e^{-\frac{(\ln r_{th} - \mu_{1})^{2}}{2\sigma_{1}^{2}}} \frac{\beta_{1}r_{th}}{\sqrt{2\pi}}$$
(18)

and for  $r_{th} \ge r_T$ :

$$\begin{split} N(r_{th}) &= \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \cdot \frac{1}{\sqrt{2\pi}\sigma_{1}r_{th}}e^{-\frac{(\ln r_{th} - \mu_{1})^{2}}{2\sigma_{1}^{2}}} \\ &\cdot \frac{\beta_{1}r_{th}}{\sqrt{2\pi}} + \frac{1 + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)}{2 + erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right) + erf\left(\frac{\ln r_{t} - \mu_{2}}{\sigma_{2}\sqrt{2}}\right)} \\ &\cdot \left(\frac{1}{2} + \frac{1}{2}erf\left(\frac{\ln r_{t} - \mu_{1}}{\sigma_{1}\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi}\sigma_{2}r_{th}}e^{-\frac{(\ln r_{th} - \mu_{2})^{2}}{2\sigma_{2}^{2}}}\frac{\beta_{2}r_{th}}{\sqrt{2\pi}} + \end{split}$$

$$+\frac{1+erf\left(\frac{\ln r_{t}-\mu_{1}}{\sigma_{1}\sqrt{2}}\right)}{2+erf\left(\frac{\ln r_{t}-\mu_{1}}{\sigma_{1}\sqrt{2}}\right)+erf\left(\frac{\ln r_{t}-\mu_{2}}{\sigma_{2}\sqrt{2}}\right)}\frac{1}{\sqrt{2\pi}\sigma_{2}r_{th}}e^{-\frac{(\ln r_{th}-\mu_{2})^{2}}{2\sigma_{2}^{2}}}\frac{\beta_{2}r_{th}}{\sqrt{2\pi}}+\frac{1+erf\left(\frac{\ln r_{t}-\mu_{1}}{\sigma_{1}\sqrt{2}}\right)}{2+erf\left(\frac{\ln r_{t}-\mu_{1}}{\sigma_{1}\sqrt{2}}\right)+erf\left(\frac{\ln r_{t}-\mu_{2}}{\sigma_{2}\sqrt{2}}\right)}\cdot\left(\frac{1}{2}+\frac{1}{2}erf\left(\frac{\ln r_{t}-\mu_{2}}{\sigma_{2}\sqrt{2}}\right)\right)\frac{1}{\sqrt{2\pi}\sigma_{th}}e^{-\frac{(\ln r_{th}-\mu_{1})^{2}}{2\sigma_{1}^{2}}}\frac{\beta_{1}r_{th}}{\sqrt{2\pi}}$$
(19)

For the channels with identical parameters it is valid for  $r_{th} < r_T$ :

$$N(r_{th}) = \left(\frac{1}{2} + \frac{1}{2} \operatorname{erf}\left(\frac{\ln r_t - \mu}{\sigma\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi}\sigma r_{th}} e^{-\frac{(\ln r_{th} - \mu)^2}{2\sigma^2}} \frac{\beta r_{th}}{\sqrt{2\pi}}$$
(20)

and for  $r_{th} \ge r_T$ :

$$N(r_{th}) = \left(\frac{3}{2} + \frac{1}{2} \operatorname{erf}\left(\frac{\ln r_t - \mu}{\sigma\sqrt{2}}\right)\right) \frac{1}{\sqrt{2\pi}\sigma r_{th}} e^{-\frac{\left(\ln r_{th} - \mu\right)^2}{2\sigma^2}} \frac{\beta r_{th}}{\sqrt{2\pi}}$$
(21)



Fig. 2. The PDF of the SSC combiner output signal and its derivative  $p_{r\dot{r}}(r,\dot{r})$  for  $r_T = 1$ ,  $\sigma = 1$ ,  $\mu = 0.5$  and  $\beta = 0.1$ 



Fig. 3. The joint PDF of the SSC combiner output signal and its derivative  $p_{r\dot{r}}(r,\dot{r})$  for  $r_T$  =1,  $\sigma$ =1.5,  $\mu$ =0.5 and  $\beta$ =0.15

## **IV. NUMERICAL RESULTS**



Fig. 4. Level crossing rate  $N(r_{th})$  for  $r_T = 1$ ,  $\sigma = 1$ ,  $\mu = 0.5$ and  $\beta = 0.1$ 



Fig. 5. Level crossing rate  $N(r_{th})$  for  $r_T = 1$ ,  $\sigma = 2$ ,  $\mu = 0.5$ and  $\beta = 0.2$ 



and  $\beta = 0.1$ 

The joint probability density functions (PDFs) of the SSC combiner output signal are shown in Figs. 2 and 3. for different values of  $r_T$ ,  $\sigma$ ,  $\mu$  and  $\beta$ . The level crossing rate curves  $N(r_{th})$ , for different parameters are given in Figs. 4 to 7.



## V. CONCLUSION

In this paper the level crossing rate (LCR) of the SSC combiner output signal is determined in the presence of lognormal fading. The results are shown graphically for different variance values and decision threshold values.

We determine LCR in order to obtain Fade Duration of SSC Combiner [7]. In our future work these system performances can be derived for correlated log-normal fading and for some other fading distributions.

#### REFERENCES

- M. S. Alouini, and M. K. Simon, "Postdetection Switched Combining- A simple Diversity Scheme With Improved BER Performance", *IEEE Trans. on Commun.*, vol. 51, No 9, Sept. 2003, pp.1591-1602.
- [2] A.A. Abu-Dayya and N. C. Beaulieu, "Analysis of switched diversity systems on generalized – fading channels", *IEEE Trans. Commun.*, vol. 42, 1994, pp. 2959-2966.
- [3] A. A. Abu-Dayya and N. C. Beaulieu, "Switched diversity on microcellular Ricean channels", *IEEE Trans. Veh. Technol.*, vol. 43, 1994, pp. 970-976.
- [4] Y. C. Ko, M. S. Alouini and M. K. Simon, "Analysis and optimization of switched diversity systems", *IEEE Trans. Veh. Technol.*, vol. 49, 2000, pp.1569-1574.
- [5] C. Tellambura, A. Annamalai and V. K. Bhargava, "Unified analysis of switched diversity systems in independent and correlated fading channels", *IEEE Trans. Commun.*, vol. 49, 1994, pp. 1955-1965.
- [6] Marvin K. Simon, Mohamed-Slim Alouni, *Digital Communication over Fading Channels*, Second Edition, Wiley Interscience, New Jersey , 2005, pp. 586.
- [7] Dragana Krstić, Petar Nikolić, Marija Matović, Ana Matović, Mihajlo Stefanović."The Outage Probability and Fade Duration of the SSC Combiner Output Signal in the Presence of Lognormal fading", accepted for *The 12th WSEAS International Conference on COMMUNICATIONS, (part of the 12th WSEAS CSCC Multiconference)*, Heraklion, Crete Island, Greece, July 22-25, 2008.