

# Parametric Analysis of Wideband CPW-fed Bow-Tie Slot Dipole

Marija Milijić<sup>1</sup> and Branka Jokanović<sup>2</sup>

**Abstract** – A uniplanar bow-tie slot dipole fed by CPW (coplanar waveguide) is proposed in this paper. Single element is designed for application in 5G frequency range 24.25-27.5 GHz. It is described how dimension parameters of bow-tie slot dipole can be designed to tune the impedance of the antenna over a wide range. The simulated results demonstrate that this structure provides the advantage of wide impedance bandwidth besides small size, low profile, easy and cheap fabrication. This antenna structure is particularly suitable for series-fed array configurations and broad-band design.

**Keywords** – Coplanar waveguide, bow-tie slot antenna, broad-band applications.

## I. INTRODUCTION

The interest for the coplanar waveguide CPW (coplanar waveguide) - fed antennas has increased significantly in recent years [1-4]. Accordingly, many antenna elements suitable for a CPW-fed configuration have been proposed to meet the necessary requirements of future 5G technologies defined as great capacities broadband service and transmission speeds. CPW-fed antennas are preferable for 5G mobile communication applications considering their low cost, light weight, small size, and ease of fabrication.

In order to set a strong foundation for the rapid advancement of the next-generation 5G networks, the International Telecommunication Union (ITU) announced the following spectrum for 5G, which includes the 24.25 – 27.5 GHz, 37 – 40.5 GHz, 66 – 76 GHz bands. Also, in the United States, the Federal Communications Commission (FCC) has announced the spectrum of approximately 11 GHz above 24 GHz for flexible, mobile and fixed wireless broadband, comprising 7 GHz of unlicensed spectrum from 64 to 71 GHz and 3.85 GHz of licensed spectrum in three bands: 27.5 to 28.35 GHz, 37 to 38.6 GHz and 38.6 to 40 GHz [5].

In the interest of meeting the requirements of wide-band applications in 5G frequency range 24.25-27.5 GHz [6,7], where high-gain and unidirectional radiation are crucial specification, this paper has investigated the CPW-fed dipoles with slots of bow-tie shape. The parameters of bow-tie slot are investigated to achieve both the desired operating frequency and impedance.

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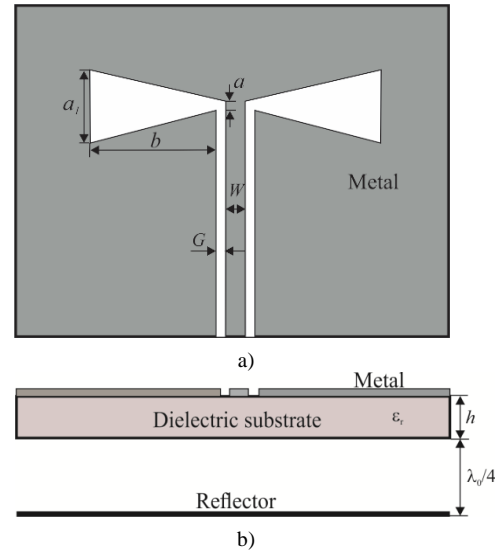


Fig. 1. CPW-fed bow-tie slot dipole: a) Top view b) Side view

## II. CPW-FED BOW-TIE SLOT DIPOLE

Proposed CPW-fed bow-tie slot dipole is depicted in Fig. 1. The radiating element and CPW feeding line are positioned as a uniplanar structure on the same side of the substrate with thickness of  $h=0.508$  mm and relative permittivity of  $\epsilon_r=2.54$ . The overall size of the substrate is  $\lambda_g \times \lambda_g$ , where  $\lambda_g=9.72$  mm is CPW line wavelength at the frequency  $f=25.875$  GHz. The bow-tie dipole has length  $b$  along which its width increases from  $a$  to  $a_1$ . The widths of the strip and gap ( $W$  and  $G$ ) of the CPW feeding line, whose impedance is around  $100 \Omega$ , are  $0.5$  mm and  $0.3$  mm, respectively [8]. The dipole is with planar reflector plate [4] at the distance  $\lambda_0/4$ , where  $\lambda_0=11.6$  mm is wavelength in air for the centre frequency of the band  $f=25.875$  GHz.

The antenna is designed in WIPL-D software [9]. The parameters of bow-tie slot dipole are investigated to achieve a wide band coverage of the antenna.

## III. SIMULATION RESULTS

First analysis was conducted to determine the advantages of bow-tie slot dipole over standard rectangular slot dipole [3]. Therefore, the initial antenna design is rectangular slot dipole with equal parameters  $a$  and  $a_1$  ( $a=a_1=0.3$  mm). The further investigation examines impedance modification for bow-tie slot whose parameter  $a$  is constant ( $a=0.3$  mm) when parameter  $a_1$  rises from  $0.5$  mm to  $1.5$  mm with step  $0.2$  mm. Simulated results are presented in Fig. 2-5 and Table I.

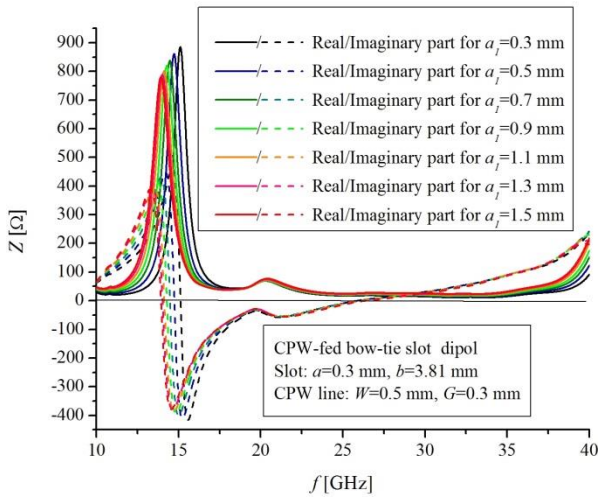


Fig. 2. Dipole's impedance versus frequency for different values of slot parameter,  $a_1$ .

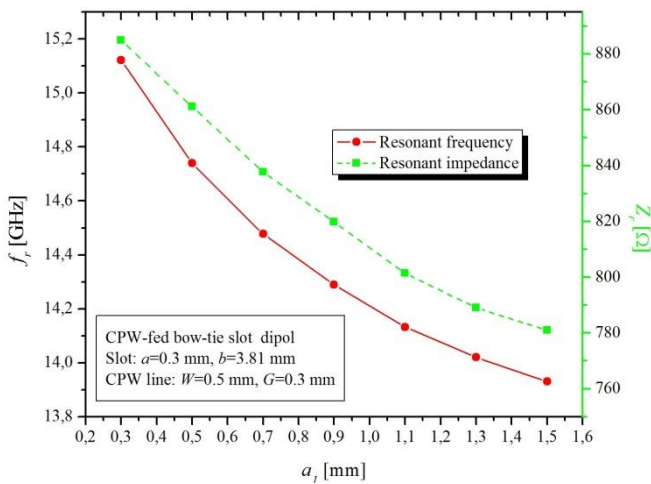


Fig. 3. Resonant frequency and impedance of CPW-fed bow-tie slot dipole for the different values of slot parameter,  $a_1$ .

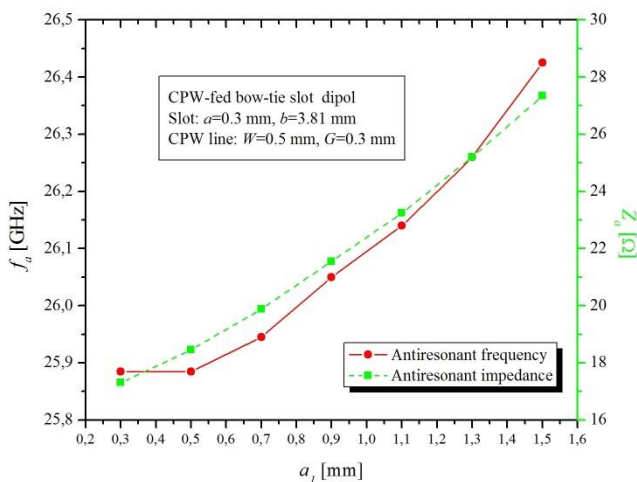


Fig. 4. Antiresonant frequency and impedance of CPW-fed bow-tie slot dipole for the different values of parameter,  $a_1$ .

Fig. 2 shows dipole's impedance versus frequency for different values of parameter  $a_1$ . It is obvious that proposed dipole has two resonances: the first resonance with high impedance and the second one or antiresonance with low impedance value.

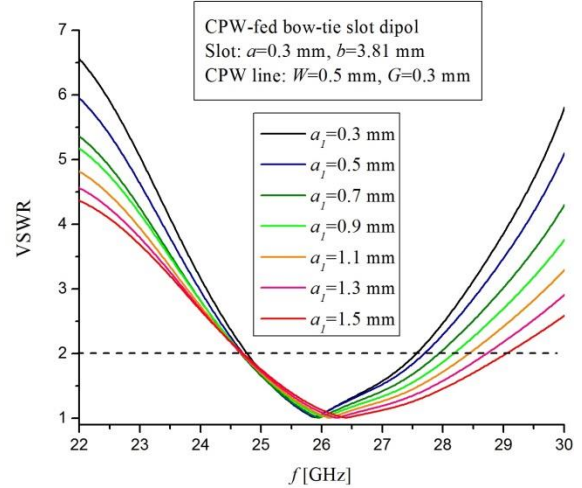


Fig. 5. Dipole's VSWR versus frequency for the different values of parameter,  $a_1$ .

TABLE I  
BANDWIDTH OF CPW-FED BOW-TIE SLOT DIPOLE FOR THE DIFFERENT VALUES OF PARAMETER  $a_1$

Slot parameter $a_1$ [mm]	Bandwidth [GHz] In respect to VSWR < 2	Relative bandwidth in respect to the center frequency [%]
0.3	24.775-27.55	10.6
0.5	24.715-27.685	11.3
0.7	24.67-27.925	12.4
0.9	24.7-28.165	13.1
1.1	24.685-28.42	14.1
1.3	24.7-28.705	15.0
1.5	24.715-29.035	16.1

It can be seen that slot impedance at the resonance varies 12.05% when parameter  $a_1$  changes from 0.3-1.5 mm, while at the second resonance the impedance has greater difference from 17  $\Omega$  to 27  $\Omega$  which makes a relative change of 45%. Considering resonance frequency in Fig. 3, it decreases for 8.2% when slot width  $a_1$  rises from 0.3 mm to 1.5 mm, while at the same time the antiresonance frequency slightly increases from 25.88 GHz to 26.4 GHz (2%) as can be seen in Fig. 4.

Since the slot impedance is very large at the resonance, it is not suitable for series-fed array of slot dipoles to work around the resonance. However, the antiresonance frequency features smaller impedance that can be observed in Fig. 4. Therefore, it can be concluded that antiresonance frequency is more suitable working band for serial connected CPW-fed bow-tie slot dipoles. Moreover, the smaller value of parameter  $a_1$  causes smaller antiresonance impedance.

On the other hand, simulated results show that as the slot width  $a_1$  increases, the bandwidth of the slot grows. Fig. 5 presents VSWR parameter of bow-tie slot dipole for different slot width  $a_1$  matching the antiresonant impedance  $Z_a$ . The numerical values from Fig. 5 are presented in Table I together with the bandwidth relative to central frequency.

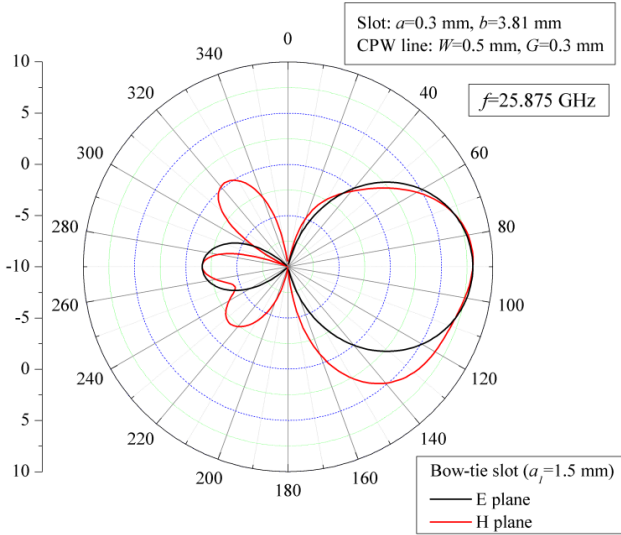


Fig. 6. Radiation pattern of the bow-tie slot dipole in E and H-plane.

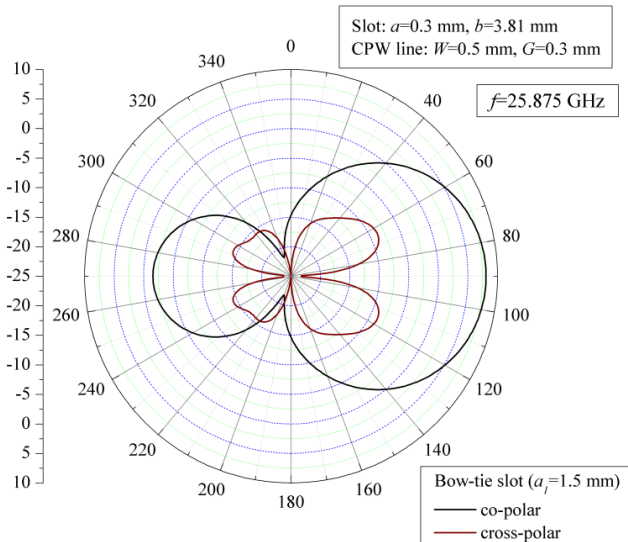


Fig. 7. Co-polar and cross-polar radiation pattern of the bow-tie slot dipole.

Analyzing previous presented results, it can be concluded that greater parameter  $a_j$  leads to bigger bandwidth. Therefore,  $a_j$  is adjusted to biggest examined value (1.5 mm) for further analysis. Fig. 6 presents simulated E-plane and H-plane radiation patterns and Fig. 7 presents simulated co- and cross-polarization of proposed bow-tie slot dipole.

Fig. 8-11 and Table II show the influence of the slot length  $b$  on resonance and antiresonance frequency bands and impedance of the proposed dipole. Since, the parameter  $b$  made a primary impact on resonance frequency band when length  $b$  changes from 3.51 mm to 4.11 mm with step 0.1 mm (Fig. 8). For this change of slot length  $b$ , resonant frequency has drop for 7.9 % (from 14.5 GHz to 13.4 GHz) while its impedance falls from 925  $\Omega$  to 650 (35 %) (Fig. 9). Still, the same change of the slot length  $b$  causes that the antiresonance frequency  $f_a$  decreases for 12.3 % (from 28.5 GHz to 25.2 GHz) while the slot impedance falls from 28  $\Omega$  to 23  $\Omega$

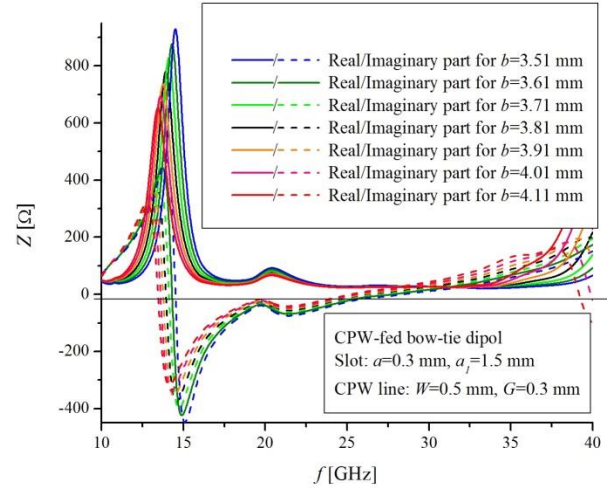


Fig. 8. Dipole's impedance versus frequency for different values of slot parameter,  $b$ .

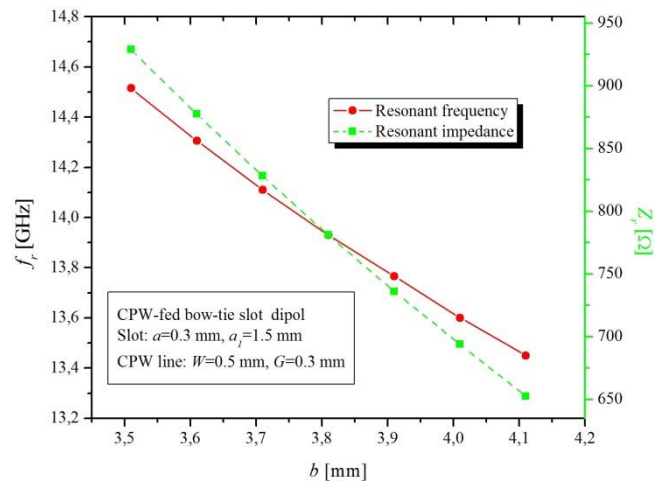


Fig. 9. Resonant frequency and impedance of CPW-fed bow-tie slot dipole for the different values of slot parameter,  $b$ .

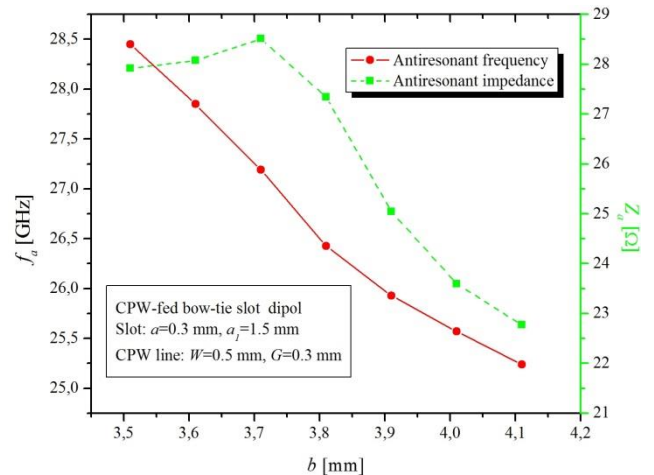


Fig. 10. Antiresonant frequency and impedance of CPW-fed bow-tie slot dipole for the different values of slot parameter,  $b$ .

(19.6 %) (Fig. 10). Furthermore, the slot impedance has calm development for both, the real and imaginary part, at the dipole second resonance (antiresonant frequency) (Fig. 8).

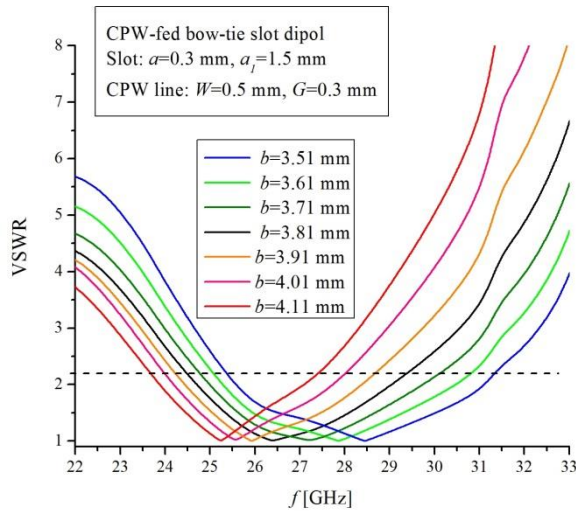


Fig. 11. Dipole's VSWR versus frequency for the different values of slot parameter,  $b$ .

TABLE II

BANDWIDTH OF CPW-FED BOW-TIE SLOT DIPOLE FOR THE DIFFERENT VALUES OF SLOT PARAMETER  $b$

Slot parameter $b$ [mm]	Bandwidth [GHz] In respect to VSWR < 2	Relative bandwidth in respect to the center frequency [%]
3.51	25.585-31.09	19.4
3.61	25.3-30.46	18.5
3.71	25-29.725	17.3
3.81	24.715-29.035	16.1
3.91	24.445-28.375	14.9
4.01	24.205-27.73	13.6
4.11	23.89-27.085	12.5

Fig. 11 shows VSWR parameter in wide frequency range matching impedance  $Z_a$  at the antiresonant frequency. Table II gives the bandwidth for the different slot length  $b$ . With respect to presented results in Fig. 11 and table II, it is obvious that the operating bandwidth of a bow-tie slot decreases when the slot length  $b$  grows.

#### IV. CONCLUSION

A CPW-fed dipole with slot of bow-tie shape is proposed for working in the frequency band 24.25-27.5 GHz intended for future 5G cellular networks. The dipole design is compact, with dimensions of only  $9.72 \times 9.72 \times 0.508 \text{ mm}^3$ . It has a simple geometry and is relatively easy to fabricate because of its single-layer metallic structure. Besides its uniplanar structure that can satisfy the requirements of low profile, the presented simulated results show that proposed dipole features great bandwidth with good impedance matching. A wider bandwidth can be achieved by correctly chosen parameters of a bow-tie slot. It is shown that both parameters  $a_1$  (bigger width of bow-tie shape) and parameter  $b$  (length of bow-tie slot) effects the bandwidth of the antenna. The demonstrated

results indicate that wider bandwidth can be achieved if the antiresonance frequencies are considered, since there is a calm variation of the impedance, for both the real and imaginary part. Furthermore, slot dipole has significantly smaller impedance at antiresonance frequency band.

Presented results have great importance in antenna array design. In order to obtain desired antenna parameters like radiation pattern, gain, side lobe suppression, etc. a large number of dipoles connected in array are required. Considering serial connection of the slot dipoles and CPW feeding line, their impedance should be small enough (around  $20 \Omega$ ) resulting in acceptable input impedance of the array. Therefore, the adjusting of slot dipole's impedance is a very important step in the beginning of antenna array design.

Further research will be directed to design of a millimeter-wave frequency scanning antenna array for radar sensors [10], which consist of a frequency modulated continuous wave radar in combination with a frequency scanning antenna.

#### ACKNOWLEDGEMENT

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#### REFERENCES

- [1] G. H. Elzwawi, M. Mantash and T. A. Denidni, "Improving the gain and directivity of CPW antenna by using a novel AMC surface," 2017 IEEE International Symposium on Antennas and Propagation, San Diego, CA, 2017, pp. 2651-2652.
- [2] A. Nešić, "Slotted Antenna Array Excited by a Coplanar Waveguide," *Electron. Lett.*, vol. 18, no. 6, pp. 275-276, 1982.
- [3] A. U. Bhohe, C. L. Holloway, M. Picket-May, R. Hall, "Wide-Band Slot Antennas With CPW Feed Lines: Hybrid and Log-Periodic Designs", *IEEE Trans. Antennas Propag.*, vol. 52, no. 10, pp. 2545-2554, October, 2004.
- [4] J. Joubert, J. C. Vardaxoglou, W. G. Whittow, J. W. Odendaal, "CPW-Fed Cavity-Backed Slot Radiator Loaded With an AMC Reflector", *IEEE Trans. Antennas Propag.*, vol. 60, no. 2, pp. 735-742, February, 2012.
- [5] M. J. Marcus, "5G and "IMT for 2020 and beyond," *IEEE Wireless Commun.*, vol. 22, no. 4, pp. 2-3, Aug. 2015
- [6] Vojislav Milosevic, Branka Jokanovic, Olga Boric-Lubecke, Victor M. Lubecke, "Key Microwave and Millimeter Wave Technologies for 5G Radio," in *Powering the Internet of Things with 5G Networks*, V. Mohanan, R. Budiatri, I. Aldmour, Eds. IGI Global, July 2017, DOI: 10.4018/978-1-5225-2799-2.
- [7] W. Hong, K. Baek, S. Ko "Millimeter-Wave 5G Antennas for Smartphones: Overview and Experimental Demonstration", *IEEE Trans. Antennas Propag.*, vol. 65, no. 12, pp. 6250 - 6261, December 2017.
- [8] M. Milijic, A. Nešić and B. Milovanović, "Study of dielectric substrate effect on modelling CPW-fed slot antenna arrays," 2017 13th International Conference TELSIS, Nis, 2017, pp. 105-108.
- [9] WIPL-D Pro v10.0, WIPL-D Team
- [10] J. Schafer, B. Gattel, H. Gulan, T. Zwick, "Integrated planar 122 GHz FMCW radar with frequency scanning antenna", *Wireless Sensors and Sensor Networks (WiSNet)*, 2018 IEEE Topical Conference on, pp. 41-43.